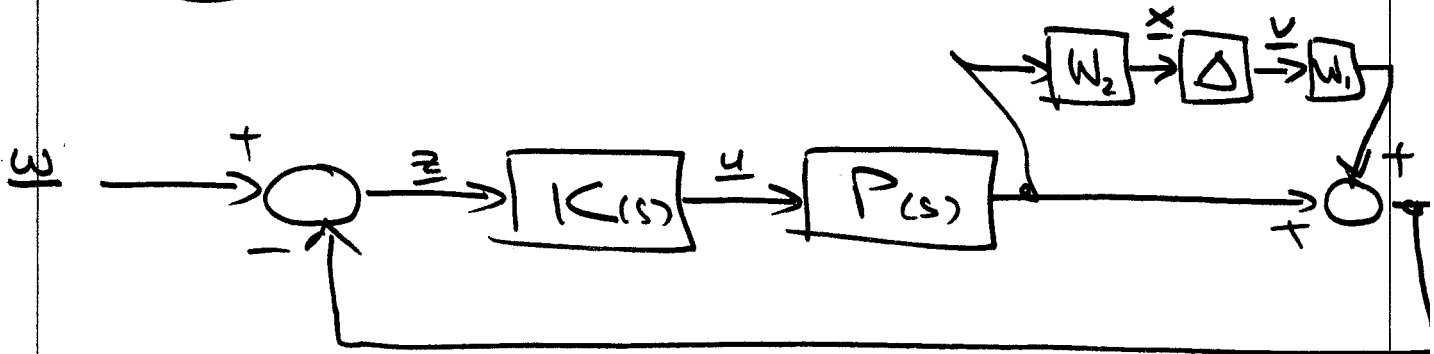


It is always possible to pull out the controllers from below and the normalized uncertainties from above.

Example:



i.e.,  $\Pi(s) = (I + W_1(s) \cdot \Delta \cdot W_2(s)) P(s)$

$$\begin{cases} \underline{x} = W_2 \cdot P \cdot \underline{u} \\ \underline{y} = W_1 \cdot \underline{v} + P \cdot \underline{u} \\ \underline{z} = \underline{w} - W_1 \cdot \underline{v} - P \cdot \underline{u} \end{cases}$$

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 42,381 50 SHEETS EYE-EASE® 5 SQUARE  
 42,382 100 SHEETS EYE-EASE® 5 SQUARE  
 42,383 200 SHEETS EYE-EASE® 5 SQUARE  
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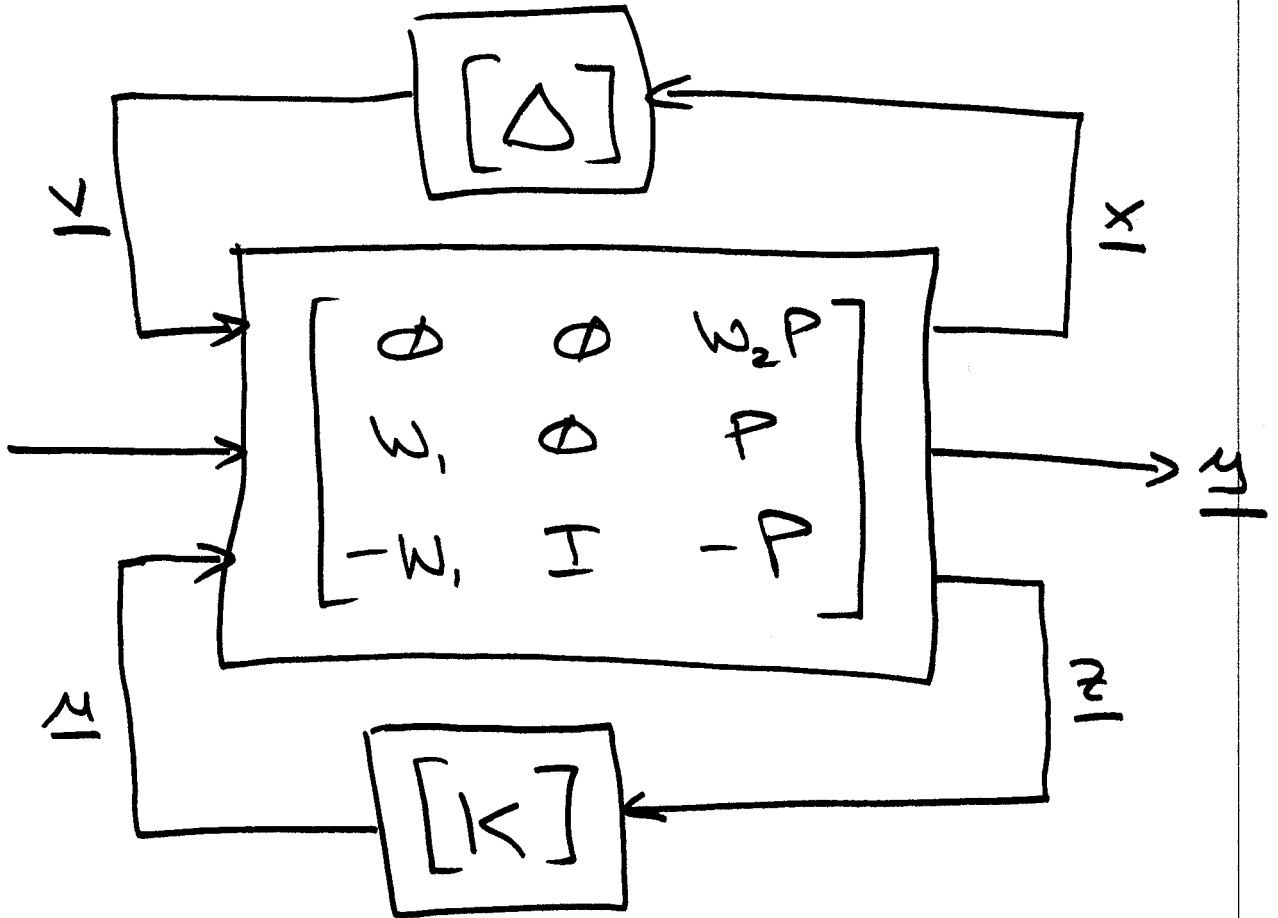
→

$$\begin{bmatrix} X \\ Y \\ W \end{bmatrix} = \begin{bmatrix} \emptyset & \emptyset & \omega_2 P \\ \omega_1 & \emptyset & P \\ -\omega_1 & I & -P \end{bmatrix} \begin{bmatrix} U \\ V \\ K \end{bmatrix}$$

$$Y = [K] W$$

$$U = [\Delta] X$$

$\hat{D}(s)$



is an LFT parametrization of the system.

$$\begin{aligned} \Rightarrow T_{wy}(s) &= \mathcal{F}_u(\mathcal{F}_\ell(\hat{P}, K), \Delta) \\ &\equiv \mathcal{F}_\ell(\mathcal{F}_u(\hat{P}, \Delta), K) \end{aligned}$$

- $\hat{P}(s)$  contains everything that is known about the plant.
- $K(s)$  contains all the controllers.
- $\Delta(s)$  contains all the normalized uncertainties about the plant:

$$\|\Delta\|_\infty = 1$$

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42-384 100 SHEETS, EYE-EASER, 5 SQUARE  
42-385 100 RECYCLED, WHITE, 5 SQUARE  
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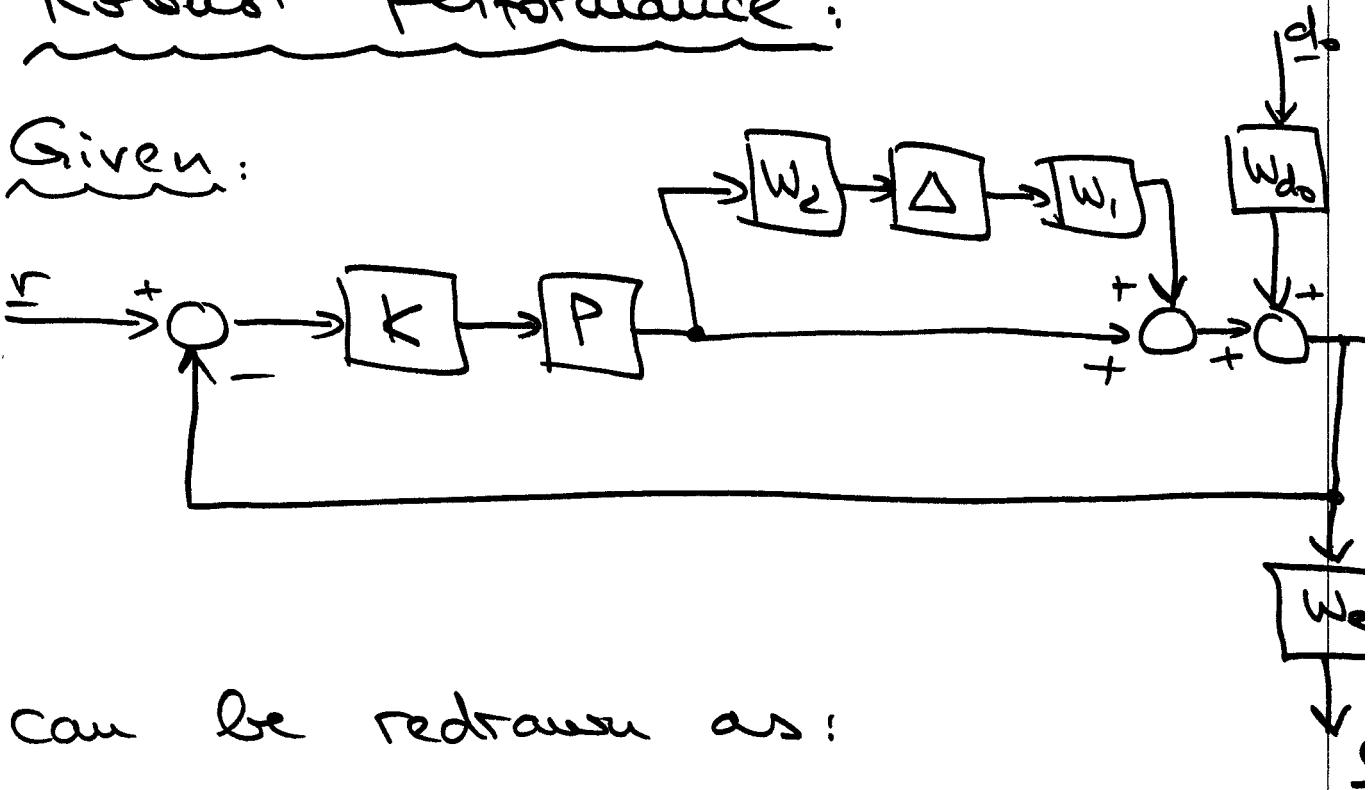
and robust stability implies:

$$\|P_{11}\|_{\infty} = \|W_2 T_0 W_1\|_{\infty} < 1$$

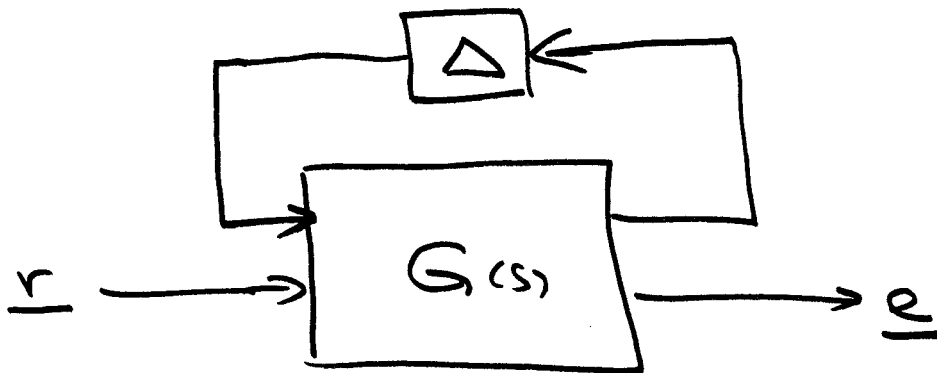
as shown earlier.

Robust performance:

Given:



can be redrawn as:



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 100 SHEETS EYE-EASE 5 SQUARE  
 42 382  
 200 SHEETS EYE-EASE 5 SQUARE  
 42 389  
 200 SHEETS EYE-EASE 5 SQUARE  
 42 392  
 100 RECYCLED WHITE 5 SQUARE  
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with:

$$G(s) = \begin{bmatrix} -W_2 T_0 W_1 & -W_2 T_0 W_{d_0} \\ W_e S_0 W_1 & W_e S_0 W_{d_0} \end{bmatrix}$$

Robust stability requires:

$$\|W_2 T_0 W_1\|_\infty < 1$$

Robust performance requires:

$$\|F_u(G, \Delta)\|_\infty < 1$$

This can be rewritten as:

$$\|G\|_\mu = \inf_{d_w \in \mathbb{R}_+} \overline{\sigma} \left( \begin{bmatrix} -W_2 T_0 W_1 & -d_w \cdot W_2 T_0 W_{d_0} \\ \frac{1}{d_w} \cdot W_e S_0 W_1 & W_e S_0 W_{d_0} \end{bmatrix} \right)$$

⇒ The robust performance problem can be reinterpreted as a robust stability problem with a second  $\Delta$  fed back from  $e$  to  $d_0$ .



The  $\| \cdot \|_u$  gives us less conservative results than the  $\| \cdot \|_2$  norm, as shall be shown shortly.

Let us assume a system with  $S$  structured sources of uncertainty:

$$\Delta_i = \delta_i I ; i = 1, \dots, S$$

and  $F$  unstructured sources of uncertainty:

$$\Delta_j ; j = 1, \dots, F$$

Pulling out all the  $\Delta_s$ , leaves us with:





not exist a  $M_{11}$  with  $\|M_{11}\|_{\infty} < 1$   
if all we assume is  $\|\Delta\|_{\infty} = 1$ .

⇒ The design may be too  
conservative.

We have thrown away  
valuable information about  
the internal structure of  
 $\Delta$ , i.e., the fact that it is  
block-diagonal.

- We can exploit this knowledge  
by reducing our requirement  
to:

$$\|M_{11}\|_{\mu} < 1$$

which may have a solution,  
even if  $\|M_{11}\|_{\infty} < 1$  does not.

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For robust performance, we simply look at  $M$  as a whole, and request:

$$\underline{\underline{\|M\|_{\mu} < 1}}$$

I already gave you an algorithm to calculate  $\|M\|_{\mu}$ . However, we still may have a BAD search problem, because  $M$  is parameterized in the set of all stabilizing controllers, either using Youla-Kucera parameterization or the technique that follows now.

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## General decomposition theorem:

Given a nominal plant  $\mathcal{P}$ :

$$\dot{\underline{x}} = A \underline{x} + B \underline{u}$$

We can find an optimal controller with:

$$\min_{\underline{u}} \int_0^{\infty} \begin{bmatrix} \underline{x} \\ \underline{u} \end{bmatrix}^* \begin{bmatrix} Q & Z \\ Z^* & R \end{bmatrix} \begin{bmatrix} \underline{x} \\ \underline{u} \end{bmatrix} dt$$

where:

$$\begin{bmatrix} Q & Z \\ Z^* & R \end{bmatrix} \succeq \Phi$$

We must assume:  $R \succ \Phi$ .

We can always normalize

$R$  to  $I$ :

$$\Rightarrow \begin{bmatrix} Q & Z \\ N^* & H \end{bmatrix} \succcurlyeq \Phi$$

If this matrix doesn't have full rank, we can decompose it as:

$$\begin{bmatrix} Q & Z \\ N^* & H \end{bmatrix} = \begin{bmatrix} C_1^* \\ D_{12}^* \end{bmatrix} \begin{bmatrix} C & D_{12} \end{bmatrix}$$

and reformulate the optimization problem as:

$$\min_{\underline{u}} \|C_1 \underline{x} + D_{12} \underline{u}\|_2$$

$\Rightarrow$  We can reinterpret this as an output equation:

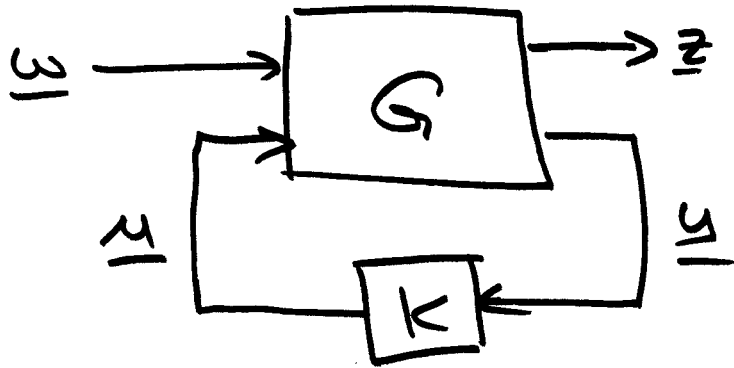
$$\left| \begin{array}{l} \dot{\underline{x}} = A \underline{x} + B_2 \underline{u} \\ \underline{y} = C_1 \underline{x} + D_{12} \underline{u} \\ \min_{\underline{u}} \|\underline{y}\|_2 \end{array} \right|$$

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42-783 500 SHEETS FULLER 9 SQUARE  
42-784 100 SHEETS FULLER 8 SQUARE  
42-785 100 SHEETS FULLER 9 SQUARE  
42-786 200 SHEETS FULLER 8 SQUARE  
42-787 200 SHEETS FULLER 9 SQUARE  
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Can be solved by Riccati equations.

Generalization:



$$G(s) = \left[ \begin{array}{c|cc} A & B_1 & B_2 \\ \hline C_1 & \emptyset & D_{12} \\ C_2 & D_{21} & \emptyset \end{array} \right]$$

⇒ We formulate the two Hamiltonians:

$$H_2 = \begin{bmatrix} A & \emptyset \\ -C_1^* & -A^* \end{bmatrix} - \begin{bmatrix} B_2 \\ -C_1^* D_{12} \end{bmatrix} \cdot \begin{bmatrix} D_{12}^* C_1 & B_2^* \end{bmatrix}$$

$$J_2 = \begin{bmatrix} A^* & \emptyset \\ -B_1 & -A \end{bmatrix} - \begin{bmatrix} C_2^* \\ -B_1 D_{21}^* \end{bmatrix} \cdot \begin{bmatrix} D_{21} B_1^* & C_2 \end{bmatrix}$$

We solve :

$$X_2 = Ric(H_2) \approx \emptyset$$

$$Y_2 = Ric(J_2) \approx \emptyset$$

$$\Rightarrow \overline{F}_2 = - (B_2^* X_2 + D_{12}^* C_1)$$

$$L_2 = - (Y_2 C_2^* + B_1 D_{21}^*)$$

$$\Rightarrow A_{F_2} = A + B_2 F_2$$

$$C_{1F_2} = C_1 + D_{12} F_2$$

---

$$A_{L_2} = A + L_2 C_2$$

$$B_{1L_2} = B_1 + L_2 D_{21}$$

---

$$A_K = A + B_2 F_2 + L_2 C_2$$

$$\Rightarrow G_C(s) = \left[ \begin{array}{c|c} A_{F_2} & I \\ \hline C_{1F_2} & \emptyset \end{array} \right];$$

$$G_L(s) = \left[ \begin{array}{c|c} A_{L_2} & B_{1L_2} \\ \hline I & \emptyset \end{array} \right];$$

$$K_{\text{nom}}(s) = \left[ \begin{array}{c|c} A_k & -L_2 \\ \hline F_2 & \phi \end{array} \right]$$

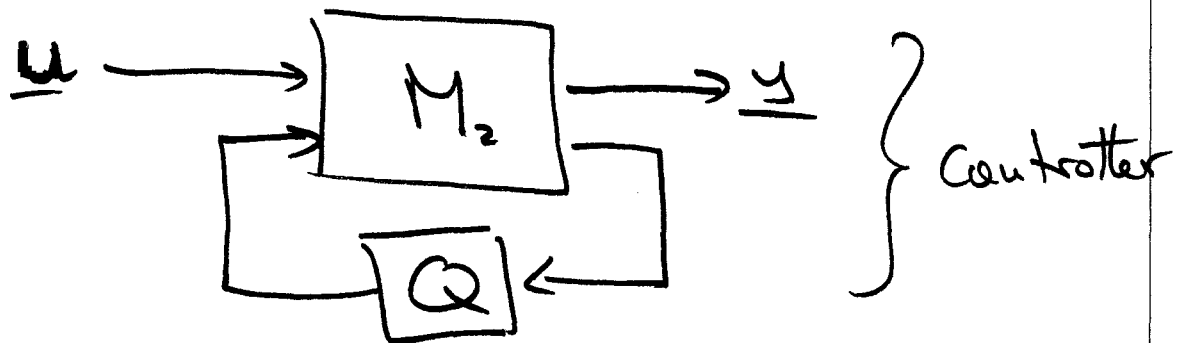
Also:

$$\begin{aligned} \|\underline{T}_{\omega_2}\|_2^2 &= \|G_c B\|_2^2 + \|F_2 G_L\|_2^2 \\ &= \|G_c L_2\|_2^2 + \|C_1 G_L\|_2^2 \end{aligned}$$

Parametrization of all stabilizing controllers:

We are interested in finding all stabilizing controllers with:

$$\|\underline{T}_{\omega_2}\|_2 < \gamma :$$



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$$M_2(s) = \left[ \begin{array}{c|cc} A_k & -L_2 & B_2 \\ \hline F_2 & \Phi & I \\ -C_2 & I & \Phi \end{array} \right]$$

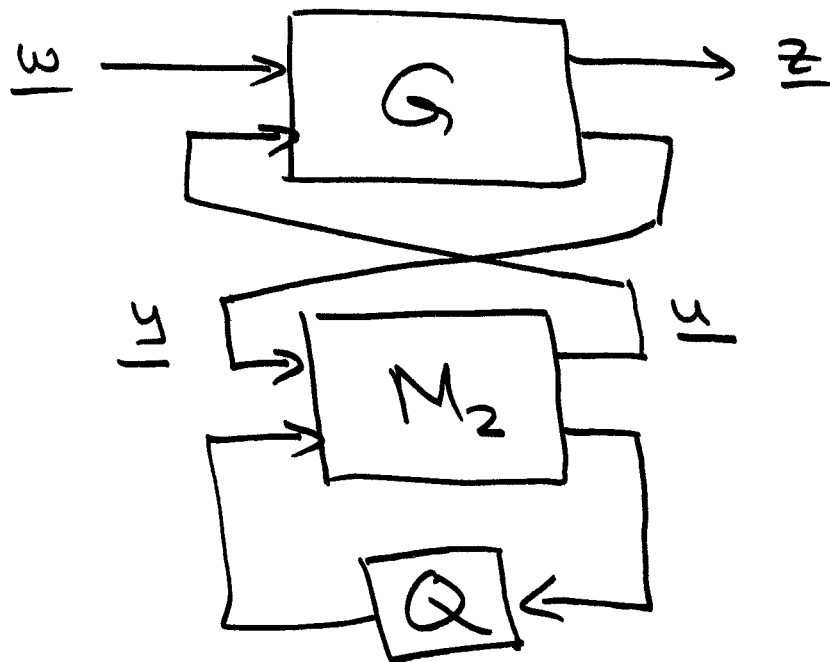
$$Q > \Phi; \quad \|Q\|_2^2 < \gamma^2 = (\|G_c B\|_2^2 + \|F_2 G_L\|_2^2)$$

is a set of suboptimal stabilizing controllers.

Clearly:  $Q = \Phi \Leftrightarrow M_2(s) = K_{nom}(s)$

---

Adding the plant:





$$\Rightarrow \left. \begin{aligned} K(s) &= F_2 (M_2(s), Q(s)) \\ T_{\underline{w}_2}(s) &= F_1 (G(s), K(s)) \end{aligned} \right\}$$

or:  $T_{\underline{w}_2}(s) = F_2 (N(s), Q(s))$

write:

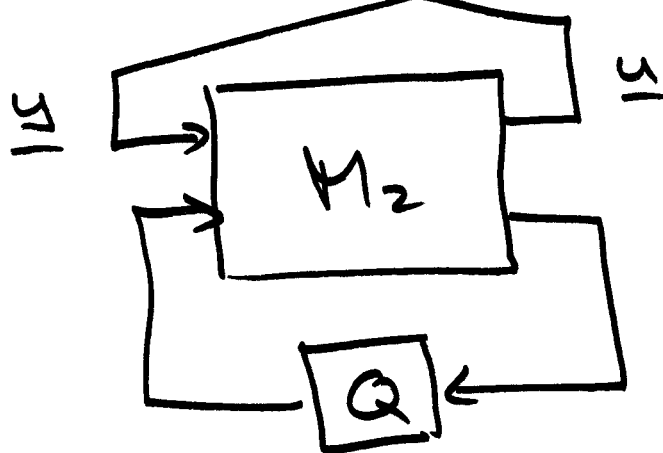
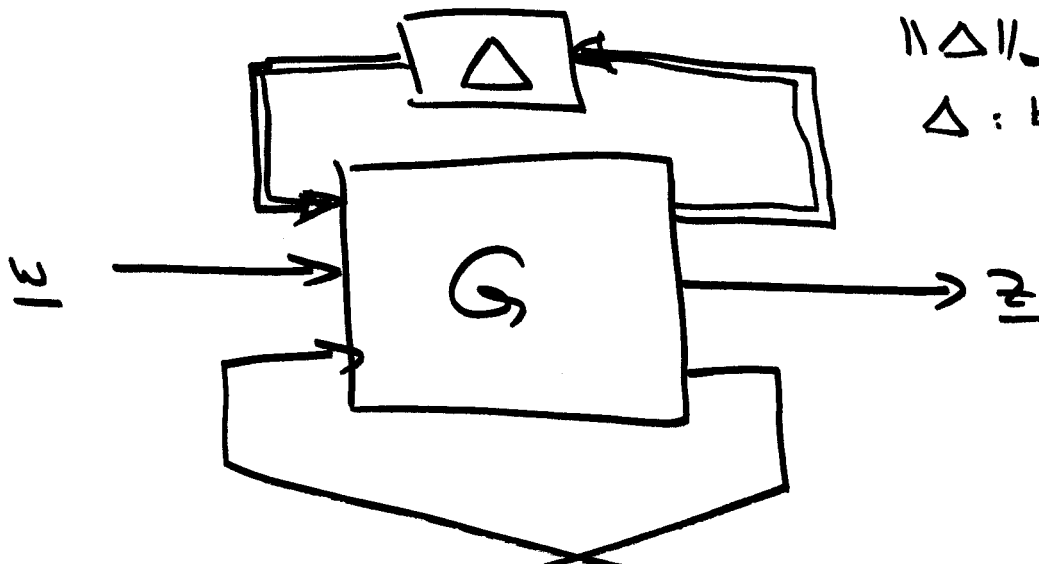
$$N(s) = \left[ \begin{array}{cc|cc} A_{F_2} & -B_2 \cdot F_2 & B_1 & B_2 \\ \phi & A_{L_2} & B_{1L_2} & \phi \\ \hline C_{1F_2} & -D_{12} \cdot F_2 & \phi & D_{12} \\ \phi & C_2 & D_{21} & \phi \end{array} \right]$$

or: let:

$$U(s) = \left[ \begin{array}{c|c} A_{F_2} & B_2 \\ \hline C_{1F_2} & D_{12} \end{array} \right]; \quad V(s) = \left[ \begin{array}{c|c} A_{L_2} & B_{1L_2} \\ \hline C_2 & D_{21} \end{array} \right]$$

$$\Rightarrow T_{\underline{w}_2}(s) = G_c B_1 - U F_2 G_c + U Q V$$

# Adding the plant uncertainty:



$$\Rightarrow K(s) = \mathcal{F}_L(M_2, Q)$$

$$M(s) = \mathcal{F}_e(G, K)$$

Choose:  $Q$  such that

$$\|M\|_\infty < 1$$

or:  $\min_Q \|M\|_\infty$

